

A.6

$$R = 4 \text{ m} \quad f = 60 \text{ MHz}$$

$$H_{11}^c \quad f_c(H_{11}) = \frac{c_0}{2\pi R / 1,8412} = 21,98 \text{ MHz}$$

$H_{11}^s$   $\hookrightarrow$  Randbedingungen  $E_{\text{tan}} = 0$  für  
 $\varphi = 0^\circ$ ;  $\varphi = 180^\circ$

$E_{01}$   $\hookrightarrow$  Randbedingungen

$$H_{21}^c \quad f_c(H_{21}) = f_c(H_{11}) \cdot 1,6588 = 36,46 \text{ MHz}$$

$H_{21}^s$   $\hookrightarrow$  RB

$$H_{01} \quad f_c(H_{01}) = f_c(H_{11}) \cdot 2,0811 = 45,74 \text{ MHz}$$

$E_{11}^c$   $\hookrightarrow$

$$E_{11}^s \quad f_c(E_{11}) = f_c(H_{11}) \cdot 2,0811 = 45,74 \text{ MHz}$$

$$H_{31}^c \quad f_c(H_{31}) = 50,15 \text{ MHz}$$

$H_{31}^s$   $\hookrightarrow$

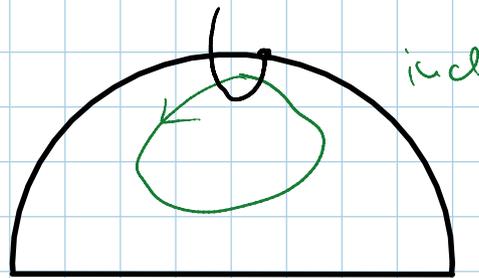
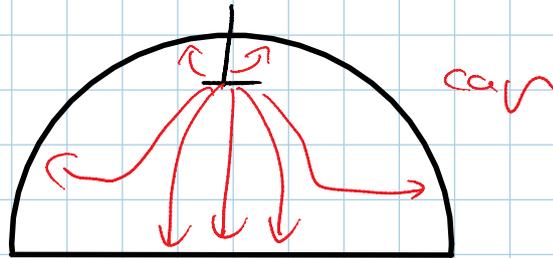
$E_{21}^c$   $\hookrightarrow$

$$E_{21}^s \quad f_c(E_{21}) = 61,3 \text{ MHz} \quad \hookrightarrow$$

Ausbreitungsfähige Moden bei  $f = 60 \text{ MHz}$

$H_{11}^c$   $H_{21}^c$   $H_{01}$   $E_{11}^s$   $H_{31}^c$

6.1.2



$\Rightarrow$   $E$ -Feld an Stelle  $\varphi = 90^\circ$   $\rho = R$   $z = 0$

am besten maximal und senkrecht zur Wand

$\Rightarrow$  nur  $H_{11}^c$   $E_{11}^s$   $H_{31}^c$  anregbar

6.1.3  $\rho = 0 \Rightarrow E = 0$

$H_{21}^c$   $H_{01}$   $H_{31}^c$

$$\begin{aligned} \underline{6.2.1)} \quad \beta_H &= \beta_{H_{11}^c} = \frac{2\pi}{\lambda_0} \sqrt{1 - \left(\frac{f_c}{f}\right)^2} \\ &= 1,1693 \text{ m}^{-1} \end{aligned}$$

$$\beta_E = \beta_{E_{11}}^S = \frac{z \cdot l}{c_0} \sqrt{1 - \left(\frac{l}{z}\right)^2}$$

$$= 0,8132 \text{ m}^{-1}$$

6.2.2)  $\vec{E}(r=0, z) = \vec{E}_{H_{11}}^c(r=0, z) + \vec{E}_{E_{11}}^s(r=0, z)$

hin  $E_0 = E_g(\varphi=90^\circ, r=R, z=0)$

$$E_\varphi(r, \varphi, z) = -\frac{\omega \mu}{\beta} H_g(r, \varphi, z)$$

$$E_g(r, \varphi, z) = \frac{\omega \mu}{\beta} H_\varphi(r, \varphi, z)$$

$$E_z(r, \varphi, z) = 0$$

mit  $H_g(r, \varphi, z) = -j \frac{z \cdot l}{\lambda_H \beta_{11}} J_1\left(1,84 \frac{r}{R}\right) \cos(\varphi) e^{-j\beta z}$

$$H_\varphi(r, \varphi, z) = j \frac{z \cdot l}{\lambda_H \cdot \beta_{11}^2 R} J_1\left(1,84 \frac{r}{R}\right) \sin(\varphi) e^{-j\beta z}$$

1.) Einspeisung bei  $r=R$   $\varphi=90^\circ$   $z=0$

$$\Rightarrow \cos(\varphi) = 0$$

$$\Rightarrow H_g = E_\varphi = 0$$

$$E_g = j \frac{\omega \mu}{\beta} \frac{z \cdot l}{\lambda_H \beta_{11}^2 R} J_1(1,84)$$

$\underbrace{\hspace{10em}}_{E_0^*}$

2.) An der Stelle  $r=0$ ,  $z=0$

$$J_1(0) = 0$$

$$\Rightarrow \Delta \varphi = \vec{E}_g = 0$$

$$\vec{E}_\varphi = j \frac{\omega \mu}{\beta} \frac{z = R}{\lambda + R_n} J_n'(0)$$

mit  $\star$

$$E_\varphi = E_0 R_n R \frac{J_n'(0)}{J_n(1.84)}$$

$$\text{mit } R_n = \frac{J_n'}{R}$$

$$E_\varphi = E_0 \frac{J_n' J_n'(0)}{J_n(1.84)} \quad J_n' = 1.841$$

$$= 1.587 E_0$$

$\vec{E}_{\text{inn}}^s$

$$\vec{E}_z = \vec{g}_z e^{-j\beta z}$$

$$E_z = \int J_m(r, \vartheta) \cos(m\varphi + \varphi_0) e^{-j\beta z}$$

$$\vec{E}_g = \vec{g}_g e^{-j\beta z}$$

$$E_g = -\frac{j\beta}{\rho^2} \int \rho \cdot J_m'(r, \vartheta) \cos(m\varphi + \varphi_0) e^{-j\beta z}$$

$$E_\varphi = \frac{j\beta}{\rho^2} \int \frac{m}{\vartheta} J_m(r, \vartheta) \sin(m\varphi + \varphi_0) e^{-j\beta z}$$

$$\vec{E}_{\text{inn}}^s: \quad m=1 \quad n=1$$

$$\rho = R \quad \vartheta = \frac{1}{R}$$

S: sinus  $z$  abh. von sinus

$$\Rightarrow \varphi_0 = 90^\circ$$

$E_{in}^s$ :

$$\Sigma_z = \underline{\Sigma} J_n \left( 3,822 \frac{s}{R} \right) \sin(\varphi) e^{-j\beta z}$$

$$E_s = -j \frac{\beta}{\rho_{in}^2} \underline{\Sigma} \rho_{in} J_n' \left( 3,832 \frac{s}{R} \right) \sin(\varphi) e^{-j\beta z}$$

$$E_\varphi = j \frac{\beta}{\rho_{in}} \underline{\Sigma} \frac{1}{s} J_n \left( 3,832 \frac{s}{R} \right) \cos(\varphi) e^{-j\beta z}$$

1.) Einspeisung an  $\varphi = 90^\circ$ ,  $s = R$ ,  $z = 0$

$$E_\varphi = 0$$

$$E_s = - \underbrace{j \frac{\beta}{\rho_{in}^2} \rho_{in} J_n' (3,832)}_{E_0}$$

2.) bei  $s = 0$ ,  $z = 0$

$$E_\varphi = 0 \quad (\text{da } J_n(0) = 0)$$

$$E_s = -j \frac{\beta}{\rho_{in}^2} \underline{\Sigma} \rho_{in} J_n'(0)$$

$$= E_0 \cdot \frac{J_n'(0)}{J_n'(3,832)} = -1,2413 E_0$$

$$\Rightarrow \vec{E}(s=0, z) = 1,582 E_0 e^{-j\beta R z} - 1,2413 E_0 e^{-j\beta z}$$

2.3) Minima von  $|\vec{E}(s=0, z)|$

$$|\vec{E}(s=0, z)| = |1,582 E_0 e^{-j\beta_H z} - 1,2413 E_0 e^{j\beta_E z}|$$

$$= 1,582 |E_0| \left| 1 - \frac{1,2413}{1,582} e^{-j(\beta_E - \beta_H)z} \right|$$

$$\Rightarrow z = \frac{u \cdot \lambda}{-(\beta_E - \beta_H)} \quad [m] \quad u \in \mathbb{N}$$

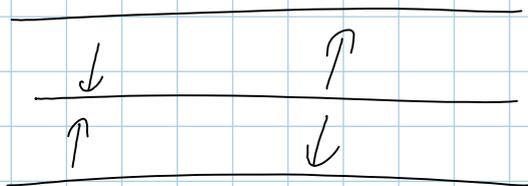
mit  $\beta_E = 0,813$  ;  $\beta_H = 1,169$

$$z = 17,63 u \quad [m] \quad u \in \mathbb{N}$$

2.4)  $V = \frac{|E_{max}|}{|E_{min}|}$  für  $s=0$

$$\frac{|E_{max}|}{|E_{min}|} = \frac{|E_0| \cdot (1,582 + 1,2413)}{|E_0| \cdot (1,582 - 1,2413)} = 8,287$$

2.5)



Feldverdracht wirkt sich nur auf den EM Wellentyp aus

→ Wellentyp kurzgeschlossen bzw.

stark gedämpft

→  $V$  wird kleiner (bzw. im Idealfall gilt  $V=1$ )

# 1.6 H08

6.1.1)  $f_{\text{CEO},1} \leq f \leq f_{\text{CEZ},1}$

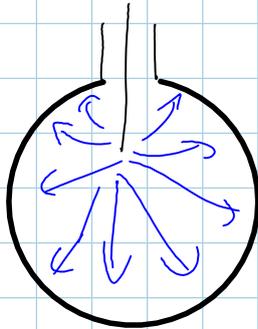
$$P_{\text{min}} = \frac{\eta_{\text{min}}}{\sqrt{2}}$$

$$\frac{c_0 \cdot 2}{z = R} \cdot 2,405 \leq f \leq \frac{c_0}{z = R} \cdot 3,054$$

$$f_{\text{CEZ},1} = P_{\text{min}} \cdot \frac{c}{2R}$$

$$3,828 \text{ GHz} \leq f \leq 6,861 \text{ GHz}$$

6.1.2)



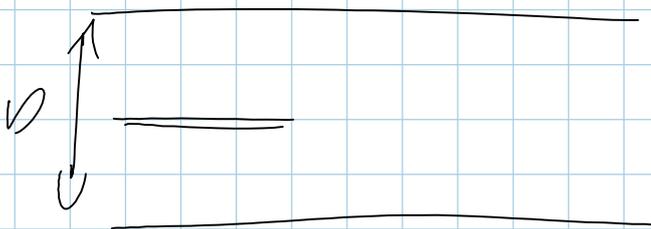
6.1.3)

$E_{0,1}$

↑ drehsymmetrisch (Zuführung für Radarsysteme)

$H_{1,1}$  SAT - Empfang

6.2)



$$d = 3 \text{ cm}$$

$$f_0 = 5 \text{ GHz}$$

6.2.1)

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### 6.2.1)

Rundhohlleiter

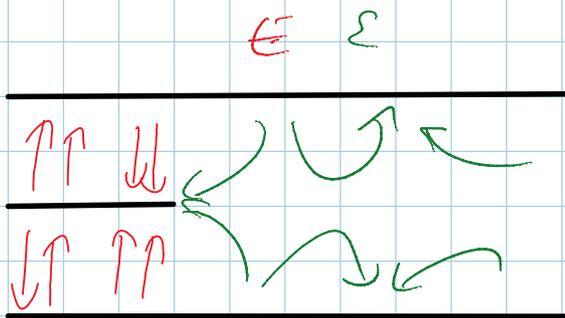
$$\begin{aligned} \rightarrow f_{c_{H_{1,1}}^{sc}} &= \nu_{0,1}^H \frac{c}{2a} = 2,93 \text{ GHz} < 5 \text{ GHz} \checkmark \\ f_{c_{E_{0,1}}} &= \nu_{0,1}^E \frac{c_0}{2a} = 3,83 \text{ GHz} < 5 \text{ GHz} \checkmark \\ f_{c_{H_{2,1}}^{sc}} &= 6,86 \text{ GHz} \checkmark \\ f_{c_{H_{0,1}}^{sc}} &= f_{c_{E_{1,1}}^{sc}} = 6,1 \text{ GHz} \checkmark \end{aligned} \left. \vphantom{\begin{aligned} \rightarrow f_{c_{H_{1,1}}^{sc}} \\ f_{c_{E_{0,1}}} \\ f_{c_{H_{2,1}}^{sc}} \\ f_{c_{H_{0,1}}^{sc}} \end{aligned}} \right\} 5 \text{ Moden}$$

Koax

$$\text{TEM} \rightarrow f_c = 0 \checkmark$$

$$\rightarrow H_{1,1} \rightarrow f_{c_{H_{1,1}}} = \frac{2c_0}{\pi(d_e + d_i)} = 6,37 \text{ GHz} \times \left. \vphantom{\frac{2c_0}{\pi(d_e + d_i)}} \right\} \Sigma^{-1}$$

### 6.2.2



die  $E_{0,1}$ -Welle (Feldlinien jeweils Rotations-symmetrisch  
transversale Feldlinien sehr ähnlich)

### 6.2.3

$$\tilde{z}_{E_{0,1}} = z_0 \sqrt{1 - \left(\frac{f_{c_{0,1}}}{f}\right)^2}$$

$$= 120 \bar{\omega} \sqrt{1 - \left(\frac{3,83}{5}\right)^2}$$

$$= 242,35 \Omega$$

$$\Rightarrow Z_{F(1,1)}^{s,c} = Z_0 \cdot \frac{1}{\sqrt{1 - \left(\frac{\beta_{c(1,1)}}{\beta}\right)^2}} = 665,24 \Omega$$

6.3)

$$d_a = 4 \text{ cm}$$

$$\frac{d_a}{d_i} = e = 2,7183$$

$$f_0 = 3 \text{ GHz}$$

$$E_{\text{max}} = 1 \text{ kV/cm}$$

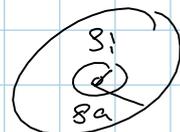
6.3.1)

$$\beta_{c(1,1)}^{k_{\text{max}}} = \frac{c_0}{\lambda_{c(1,1)}} = \frac{2 c_0}{(d_{\text{side}}) \bar{\omega}} \stackrel{!}{=} \beta_{c(1,1)}^{\text{Hohlk.}} = \frac{c_0}{2 \bar{\omega} R} \quad \bar{\omega} \text{ cm}^{-1}$$

$$D = 8,56 \text{ cm}$$

6.3.2)

$$\vec{E}_r = \frac{U}{r \ln \frac{\rho_a}{\rho_i}}$$



$$P = \frac{1}{2} \iint_A \vec{E}_r \times \vec{H}_\phi^* dA$$

$$= \frac{1}{2} \int_{\rho_i}^{\rho_a} \int_0^{2\pi} |E|^2 \cdot \dots \cdot \rho \, d\rho \, d\phi$$

$$= \frac{1}{2} \int_0^{g_2} \int_0^{2\pi} \frac{|E|^2}{z=0} g \, d\beta \, d\varphi = \frac{1}{2} \int_{g_i}^{g_2} \int_0^{2\pi} \frac{|U|^2}{g \cdot \ln^2 \frac{g_2}{g_i}} g \, d\beta \, d\varphi$$

mit

$$\Rightarrow |E_{\max}| = \frac{|U|}{g_i \ln \frac{g_2}{g_i}}$$

$$P = \frac{1}{2} \int_{g_i}^{g_2} \int_0^{2\pi} \frac{|E_{\max}|^2 g_i^2 \ln^2 \frac{g_2}{g_i}}{g \cdot \ln^2 \frac{g_2}{g_i}} d\beta \, d\varphi$$

$$= \frac{1}{2} \int_{g_i}^{g_2} \int_0^{2\pi} \frac{|E_{\max}|^2 g_i^2}{g} d\beta \, d\varphi = \frac{|E_{\max}|^2 g_i^2}{z=0} \ln \frac{g_2}{g_i}$$

$$= 4,51 \text{ kW}$$